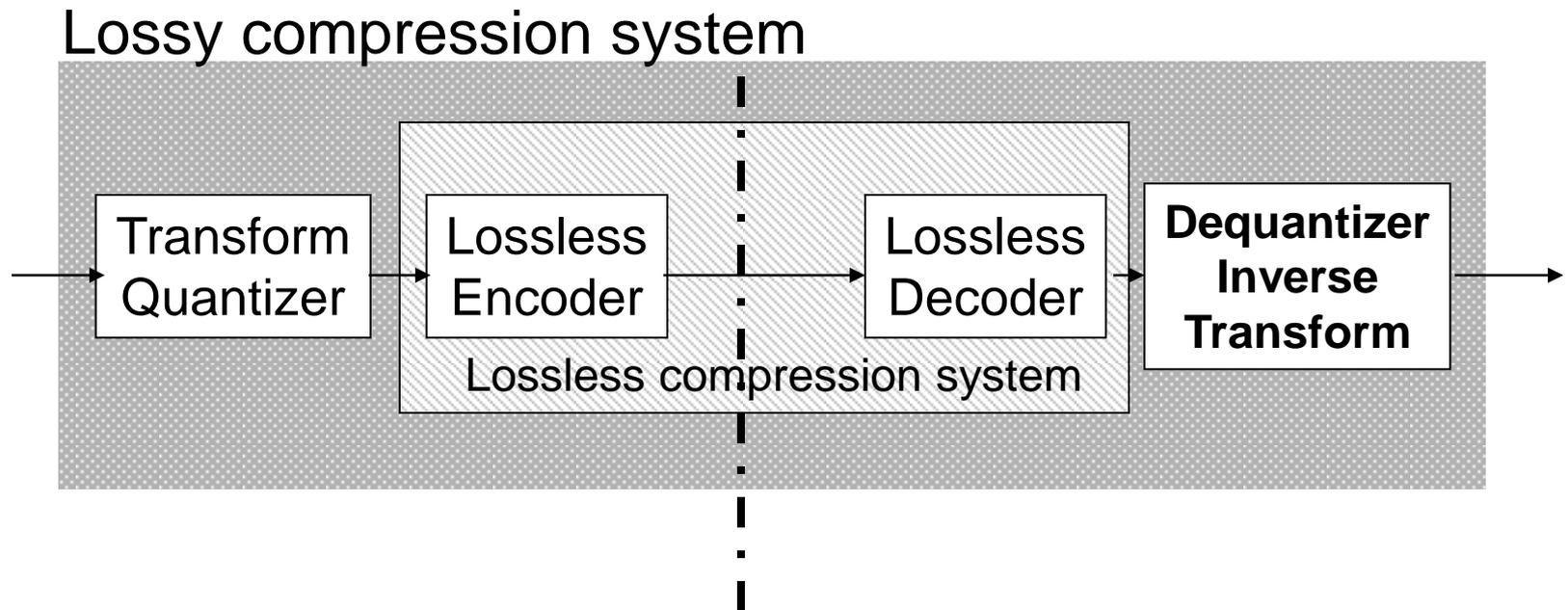


# Lossless compression in lossy compression systems

- Almost every lossy compression system contains a lossless compression system



- We discuss the basics of lossless compression first, then move on to lossy compression



# Topics in lossless compression

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- Binary decision trees and variable length coding
- Entropy and bit-rate
- Prefix codes, Huffman codes, Golomb codes
  
- Joint entropy, conditional entropy, sources with memory
- Fax compression standards
- Arithmetic coding



# Example: 20 Questions

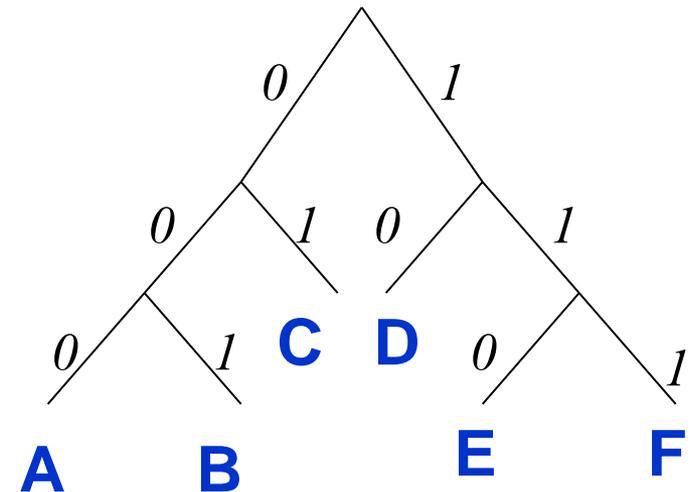
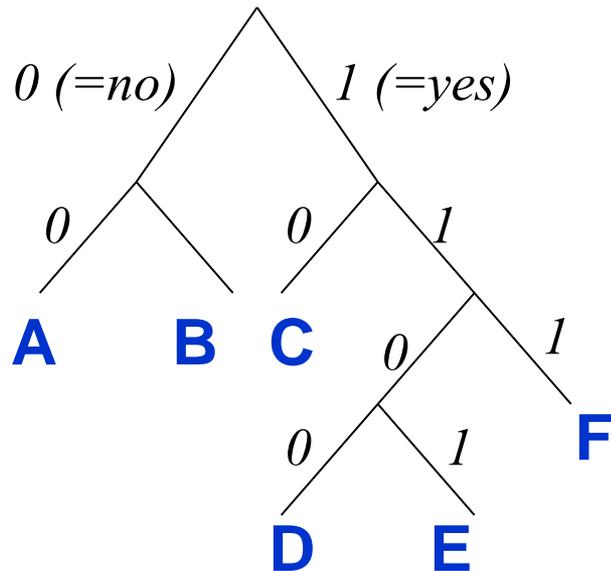
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- *Alice* thinks of an outcome (from a finite set), but does not disclose her selection.
- *Bob* asks a series of yes/no questions to uniquely determine the outcome chosen. The goal of the game is to ask as few questions as possible on average.
- Our goal: Design the best strategy for *Bob*.



# Example: 20 Questions (cont.)

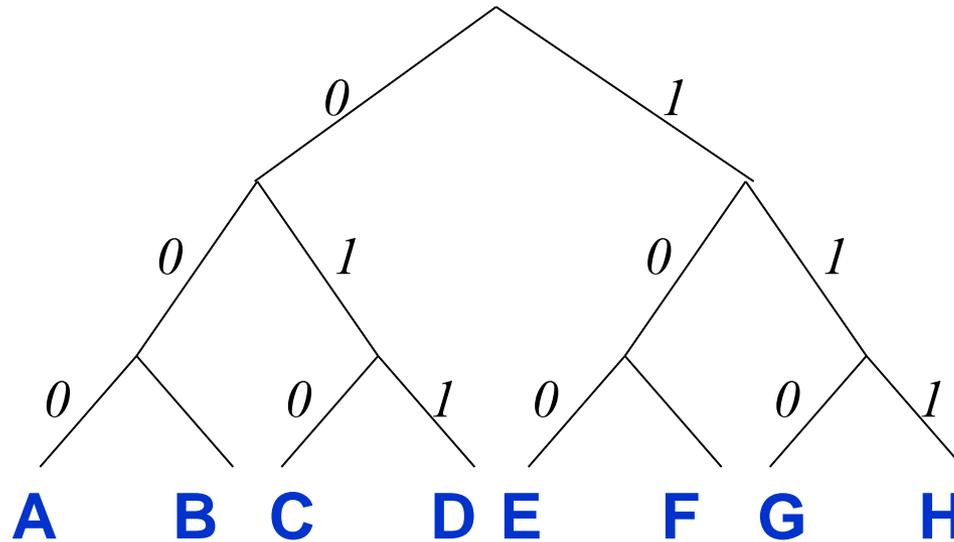
- Which strategy is better?



- Observation: The collection of questions and answers yield a binary code for each outcome.



# Fixed length codes



- Average description length for  $K$  outcomes  $l_{av} = \log_2 K$
- Optimum for equally likely outcomes
- Verify by modifying tree



# Variable length codes

- If outcomes are NOT equally probable:
  - Use shorter descriptions for likely outcomes
  - Use longer descriptions for less likely outcomes
- Intuition:
  - Optimum balanced code trees, i.e., with equally likely outcomes, can be pruned to yield unbalanced trees with unequal probabilities.
  - The unbalanced code trees such obtained are also optimum.
  - Hence, an outcome of probability  $p$  should require about

$$\log_2 \left( \frac{1}{p} \right) \text{ bits}$$



# Entropy of a random variable

- Consider a discrete, finite-alphabet random variable  $X$

$$\text{Alphabet } \mathcal{A}_X = \{\alpha_0, \alpha_1, \alpha_2, \dots, \alpha_{K-1}\}$$

$$\text{PMF } f_X(x) = P(X = x) \text{ for each } x \in \mathcal{A}_X$$

- **Information** associated with the event  $X=x$

$$h_X(x) = -\log_2 f_X(x)$$

- **Entropy of  $X$**  is the expected value of that information

$$H(X) = E[h_X(X)] = -\sum_{x \in \mathcal{A}_X} f_X(x) \log_2 f_X(x)$$

- Unit: bits



# Information and entropy: properties

- Information  $h_X(x) \geq 0$
- Information  $h_X(x)$  strictly increases with decreasing probability  $f_X(x)$
- Boundedness of entropy

$$0 \leq H(X) \leq \log_2 (\|\mathcal{A}_X\|)$$

Equality if only one outcome can occur

Equality if all outcomes are equally likely

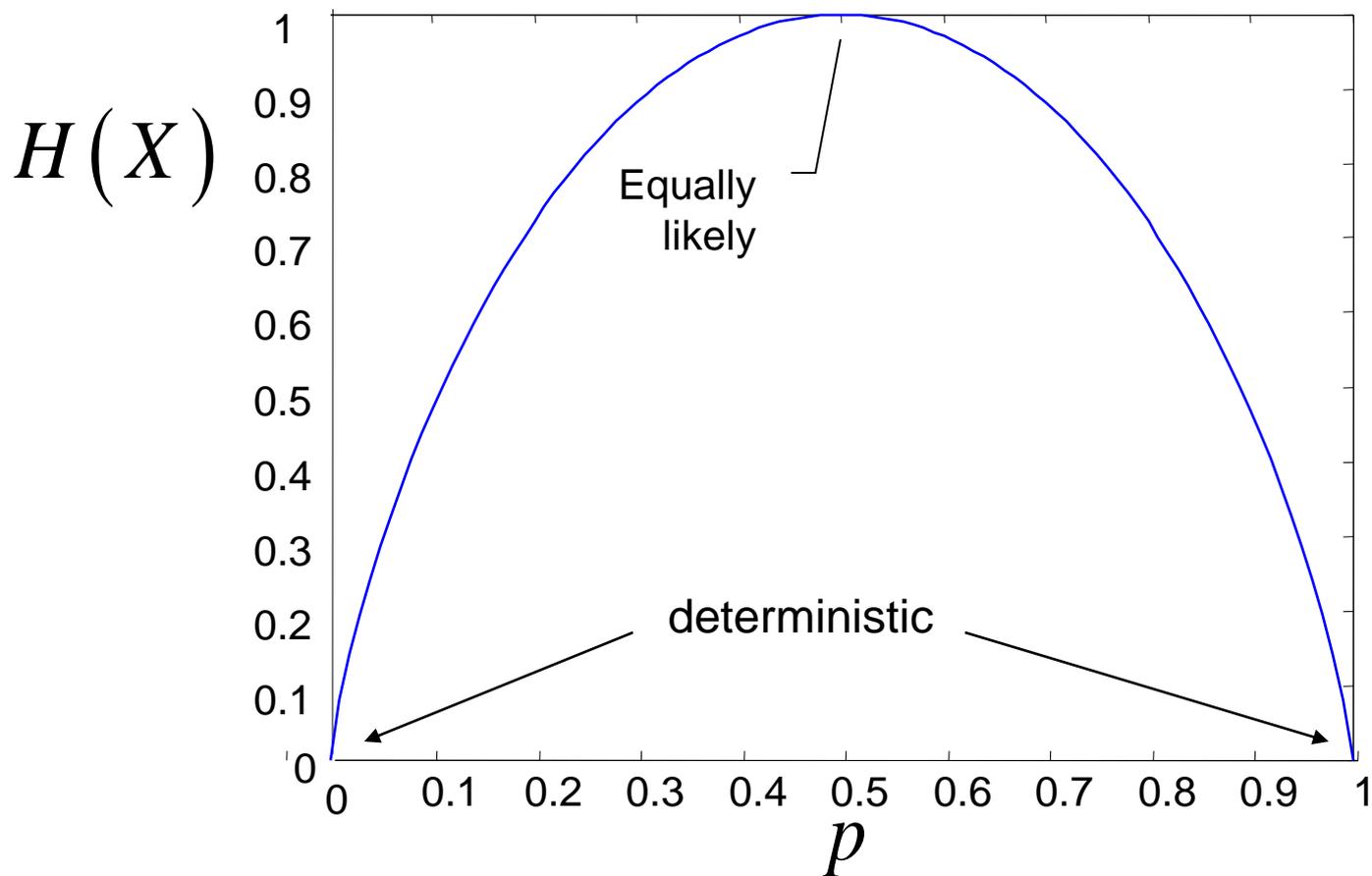
- Very likely and very unlikely events do not substantially change entropy

$$-p \log_2 p \rightarrow 0 \quad \text{for } p \rightarrow 0 \text{ or } p \rightarrow 1$$



# Example: Binary random variable

$$H(X) = -p \log_2 p - (1-p) \log_2 (1-p)$$



# Entropy and bit-rate

- Consider IID random process  $\{X_n\}$  (or “source”) where each sample  $X_n$  (or “symbol”) possesses identical entropy  $H(X)$
- $H(X)$  is called “entropy rate” of the random process.
- Noiseless Source Coding Theorem [*Shannon, 1948*]
  - The entropy  $H(X)$  is a lower bound for the average word length  $R$  of a decodable variable-length code for the symbols.
  - Conversely, the average word length  $R$  can approach  $H(X)$ , if sufficiently large blocks of symbols are encoded jointly.
- Redundancy of a code:

$$\rho = R - H(X) \geq 0$$



# Variable length codes

- Given IID random process  $\{X_n\}$  with alphabet  $\mathcal{A}_X$  and PMF  $f_X(x)$
- Task: assign a distinct code word,  $c_x$ , to each element,  $x \in \mathcal{A}_X$ , where  $c_x$  is a string of  $\|c_x\|$  bits, such that each symbol  $x_n$  can be determined, even if the codewords  $c_{x_n}$  are directly concatenated in a bitstream
- Codes with the above property are said to be “uniquely decodable.”
- Prefix codes
  - No code word is a prefix of any other codeword
  - Uniquely decodable, symbol by symbol, in natural order  $0, 1, 2, \dots, n, \dots$





# Unique decodability: McMillan and Kraft conditions

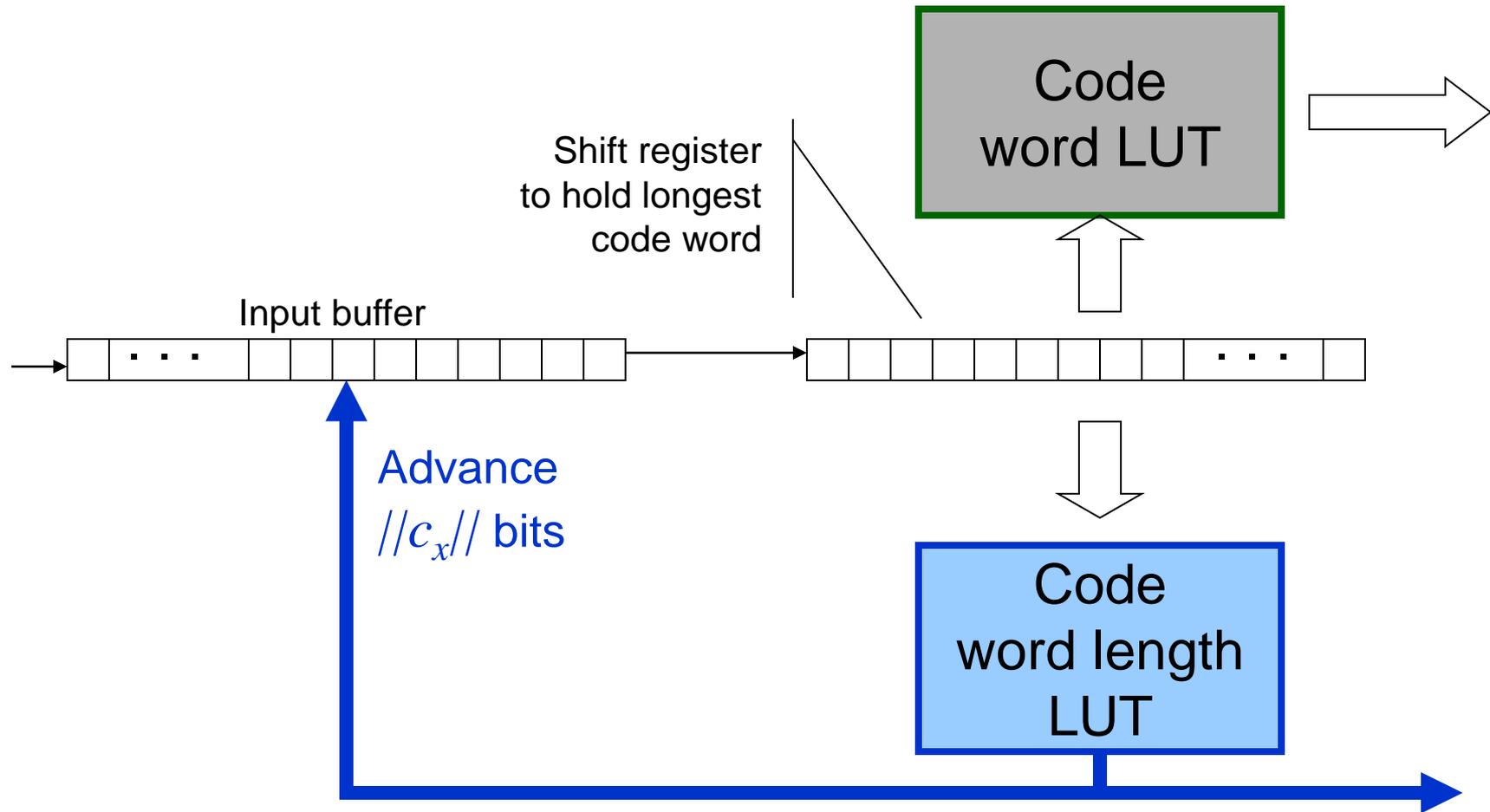
- Necessary condition for unique decodability [*McMillan*]

$$\sum_{x \in \mathcal{A}_X} 2^{-\|c_x\|} \leq 1$$

- Given a set of code word lengths  $\|c_x\|$  satisfying McMillan condition, a corresponding prefix code always exists [*Kraft*]
  - Hence, McMillan inequality is both necessary and sufficient.
  - Also known as Kraft inequality or Kraft-McMillan inequality.
  - No loss by only considering prefix codes.
  - Prefix code is not unique.

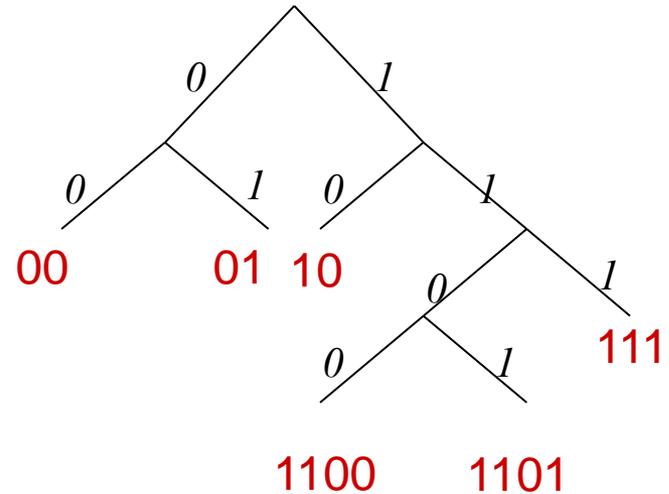


# Prefix Decoder



# Binary trees and prefix codes

- Any binary tree can be converted into a prefix code by traversing the tree from root to leaves.
- Any prefix code corresponding to a binary tree meets McMillan condition with equality



$$3 \cdot 2^{-2} + 2 \cdot 2^{-4} + 2^{-3} = 1$$

$$\sum_{x \in \mathcal{A}_X} 2^{-\|c_x\|} = 1$$





# Instantaneous variable length encoding without redundancy

- A code without redundancy, i.e.

$$R = H(X)$$

requires all individual code word lengths

$$l_{\alpha_k} = -\log_2 f_X(\alpha_k)$$

- All probabilities would have to be binary fractions:

$$f_X(\alpha_k) = 2^{-l_{\alpha_k}}$$

Example

$\alpha_i$	$P(\alpha_i)$	redundant code	optimum code
$\alpha_0$	0.500	00	0
$\alpha_1$	0.250	01	10
$\alpha_2$	0.125	10	110
$\alpha_3$	0.125	11	111

$$H(X) = 1.75 \text{ bits}$$

$$R = 1.75 \text{ bits}$$

$$\rho = 0$$



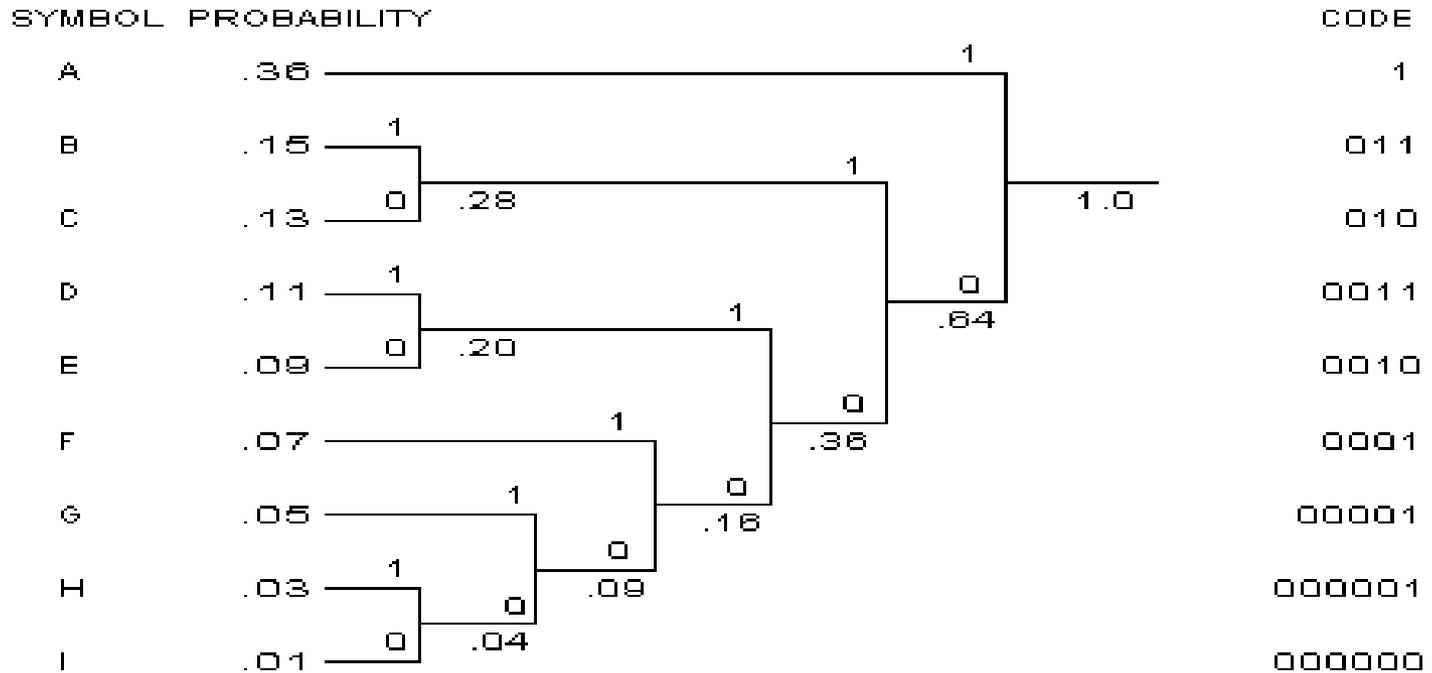
# Huffman Code

- Design algorithm for variable length codes proposed by Huffman (1952) always finds a code with minimum redundancy.
- Obtain code tree as follows:

- 1 Pick the two symbols with lowest probabilities and merge them into a new auxiliary symbol.
- 2 Calculate the probability of the auxiliary symbol.
- 3 If more than one symbol remains, repeat steps 1 and 2 for the new auxiliary alphabet.
- 4 Convert the code tree into a prefix code.



# Huffman Code - Example



Fixed length coding:	$R_{fixed} = 4$ bits/symbol
Huffman code:	$R_{Huffman} = 2.77$ bits/symbol
Entropy	$H(X) = 2.69$ bits/symbol
Redundancy of the Huffman code:	$\rho = 0.08$ bits/symbol



# Redundancy of prefix code for general distribution

- Huffman code redundancy  $0 \leq \rho < 1$  bit/symbol
- Theorem: For any distribution  $f_X$ , a prefix code can be found, whose rate  $R$  satisfies

$$H(X) \leq R < H(X) + 1$$

- Proof

- Left hand inequality: Shannon's noiseless coding theorem
- Right hand inequality:

Choose code word lengths  $\|c_x\| = \lceil -\log_2 f_X(x) \rceil$

$$\begin{aligned} \text{Resulting rate } R &= \sum_{x \in \mathcal{A}_X} f_X(x) \lceil -\log_2 f_X(x) \rceil \\ &< \sum_{x \in \mathcal{A}_X} f_X(x) (1 - \log_2 f_X(x)) \\ &= H(X) + 1 \end{aligned}$$



# Vector Huffman coding

- Huffman coding very inefficient for  $H(X) \ll 1$  bit/symbol
- Remedy:
  - Combine  $m$  successive symbols to a new “block-symbol”
  - Huffman code for block-symbols
  - Redundancy

$$H(X) \leq R < H(X) + \frac{1}{m}$$

- Can also be used to exploit statistical dependencies between successive symbols
- Disadvantage: exponentially growing alphabet size  $\|\mathcal{A}_X\|^m$



# Truncated Huffman Coding

- Idea: reduce size of Huffman code table and maximum Huffman code word length by Huffman-coding only the most probable symbols.
  - Combine  $J$  least probable symbols of an alphabet of size  $K$  into an auxiliary symbol  $ESC$
  - Use Huffman code for alphabet consisting of remaining  $K-J$  most probable symbols and the symbol  $ESC$
  - If  $ESC$  symbol is encoded, append  $\lceil \log_2(J) \rceil$  bits to specify exact symbol from the full alphabet
- Results in increased average code word length – trade off complexity and efficiency by choosing  $J$



# Adaptive Huffman Coding

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- Use, if source statistics are not known ahead of time
- Forward adaptation
  - Measure source statistics at encoder by analyzing entire data
  - Transmit Huffman code table ahead of compressed bit-stream
  - JPEG uses this concept (even though often default tables are transmitted)
- Backward adaptation
  - Measure source statistics both at encoder and decoder, using the same previously decoded data
  - Regularly generate identical Huffman code tables at transmitter and receiver
  - Saves overhead of forward adaptation, but usually poorer code tables, since based on past observations
  - Generally avoided due to computational burden at decoder



# Unary coding

- “Geometric” source

$$\text{Alphabet } \mathcal{A}_x = \{0, 1, \dots\} = \mathbb{Z}_+ \quad \text{PMF } f_x(x) = 2^{-(x+1)}, \quad x \geq 0$$

- Optimal prefix code with redundancy  $\rho=0$  is “unary” code (“comma code”)

$$c_0 = "1" \quad c_1 = "01" \quad c_2 = "001" \quad c_3 = "0001" \quad \dots$$

- Consider geometric source with faster decay

$$\text{PMF } f_x(x) = (1 - \beta) \beta^x, \text{ with } 0 \leq \beta < \frac{1}{2}; \quad x \geq 0$$

- Unary code is still optimum prefix code (i.e., Huffman code), but not redundancy-free



# Golomb coding

- For geometric source with slower decay

$$\text{PMF } f_X(x) = (1 - \beta)\beta^x, \text{ with } \frac{1}{2} < \beta < 1; \quad x \geq 0$$

- Idea: Express each  $x$  as

$$x = mx_q + x_r \quad \text{with} \quad x_q = \left\lfloor \frac{x}{m} \right\rfloor \quad \text{and} \quad x_r = x \bmod m$$

- Distribution of new random variables

$$f_{X_q}(x_q) = \sum_{i=0}^{m-1} f_X(mx_q + i) = \beta^{mx_q} \sum_{i=0}^{m-1} f_X(i)$$

$$f_{X_r}(x_r) = \frac{1 - \beta}{1 - \beta^m} \beta^{x_r} \quad \text{for } 0 \leq x_r < m$$

$X_q$  and  $X_r$  statistically independent.



# Golomb coding (cont.)

## ■ Golomb coding

- Choose integer divisor  $\beta^m \approx \frac{1}{2}$
- Encode  $x_q$  optimally by unary code
- Encode  $x_r$  by a modified binary code, using code word lengths

$$k_a = \lceil \log_2 m \rceil$$

$$k_b = \lfloor \log_2 m \rfloor$$

- Concatenate bits for  $x_q$  and  $x_r$
- 
- In practice,  $m=2^k$  is often used, so  $x_r$  can be encoded by constant code word length  $\log_2 m$



# Golomb code examples

Unary  
Code

1  
01  
001  
0001  
00001  
000001  
0000001  
00000001  
000000001  
.  
.  
.

Unary  
Code

10  
11  
010  
011  
0010  
0011  
00010  
00011  
000010  
000011  
0000010  
0000011  
00000010  
00000011  
.  
.  
.

Constant  
length code

$m=2$

Unary  
Code

100  
101  
110  
111  
0100  
0101  
0110  
0111  
00100  
00101  
00110  
00111  
000100  
000101  
000110  
000111  
.  
.  
.

Constant  
length code

$m=4$



# Golomb parameter estimation

- Expected value for geometric distribution

$$E[X] = \sum_{x=0}^{\infty} (1-\beta) x \beta^x = \frac{\beta}{1-\beta} \quad \rightarrow \quad \beta = \frac{E[X]}{1+E[X]}$$

- Approximation for  $E[X] \gg 1$

$$\beta^m = \frac{(E[X])^m}{(1+E[X])^m} \approx 1 - \frac{m}{E[X]} \approx \frac{1}{2}$$

$$m = 2^k \approx \frac{1}{2} E[X]$$

$$k = \max \left\{ 0, \left\lceil \log_2 \left( \frac{1}{2} E[X] \right) \right\rceil \right\}$$

Rule for optimum performance of Golomb code

Reasonable setting, even if  $E[X] \gg 1$  does not hold



# Adaptive Golomb coder (JPEG-LS)

